Encoding/Decoding Strategies and Rate Regions for Cooperative Multiple Access Channels

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- Interference is information.
- Some versions of all transmitted signals are received by all nodes.
- User cooperation: exploit overheard information to jointly design encoding, transmit, routing policies.
- Building block towards the analysis of larger networks.







Motivation



Motivation



Optimum Power Allocation for the Two User Cooperative MAC



 $Y_0 = h_{10}X_1 + h_{20}X_2 + n_0$ $Y_1 = h_{21}X_2 + n_1$ $Y_2 = h_{12}X_1 + n_2$

• Joint work with Sennur Ulukus

MAC with Generalized Feedback

- Gaussian MAC with cooperating encoders [Sendonaris, Erkip, Aazhang]
 - Special case of MAC with generalized feedback [Willems, van der Meulen, Schalkwijk]
- An achievable rate region is obtained by employing
 - Block Markov superposition encoding
 - * Inject high rate fresh information to be resolved with the help of upcoming blocks.
 - * Send resolution information for previous blocks.
 - Backward decoding
 - * After receiving all blocks, decode the resolution information in the last block.
 - * Using previously decoded resolution information, sequentially decode earlier blocks.



- Build common information (X_{12}, X_{21})
- Cooperatively send (U)
- Inject new information (X_{10}, X_{20})

$$X_{1} = \sqrt{p_{10}}X_{10} + \sqrt{p_{12}}X_{12} + \sqrt{p_{u1}}U$$
$$X_{2} = \sqrt{p_{20}}X_{20} + \sqrt{p_{21}}X_{21} + \sqrt{p_{u2}}U$$

- Amplitude of the each channel's gain is assumed to be known at the corresponding receiver.
- Phases of all channel gains are assumed known at the receiver and the transmitters
 - Coherent combining.



- Build common information (X_{12}, X_{21})
- Cooperatively send (U)
- Inject new information (X_{10}, X_{20})

$$X_{1} = \sqrt{p_{10}(\mathbf{h})} X_{10} + \sqrt{p_{12}(\mathbf{h})} X_{12} + \sqrt{p_{u1}(\mathbf{h})} U$$
$$X_{2} = \sqrt{p_{20}(\mathbf{h})} X_{20} + \sqrt{p_{21}(\mathbf{h})} X_{21} + \sqrt{p_{u2}(\mathbf{h})} U$$

- Complete channel state information at the transmitters and the receiver.
- Transmitted codewords can be modulated by channel adaptive power levels
 - Opportunistic cooperation and transmission use available average power efficiently.



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$$X_2 = \sqrt{p_{20}(\mathbf{h})} X_{20}$$

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- Complete channel state information at the transmitters and the receiver.
- Transmitted codewords can be modulated by channel adaptive power levels
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Achievable Region of Rates with Power Control

• Union over all valid policies $E[p_{i0}(\mathbf{h}) + p_{ij}(\mathbf{h}) + p_{U_i}(\mathbf{h})] \leq \bar{p}_i$ of pairs $\{R_1, R_2\}$ that satisfy

$$\begin{split} R_{1} &< E\left[\log\left(1 + \frac{h_{12}p_{12}(\mathbf{h})}{h_{12}p_{10}(\mathbf{h}) + \sigma_{2}^{2}}\right) + \log\left(1 + \frac{h_{10}p_{10}(\mathbf{h})}{\sigma_{0}^{2}}\right)\right] \\ R_{2} &< E\left[\log\left(1 + \frac{h_{21}p_{21}(\mathbf{h})}{h_{21}p_{20}(\mathbf{h}) + \sigma_{1}^{2}}\right) + \log\left(1 + \frac{h_{20}p_{20}(\mathbf{h})}{\sigma_{0}^{2}}\right)\right] \\ R_{1} + R_{2} &< \min\left\{E\left[\log\left(1 + \frac{h_{10}p_{1}(\mathbf{h}) + h_{20}p_{2}(\mathbf{h}) + 2\sqrt{h_{10}h_{20}p_{U_{1}}(\mathbf{h})p_{U_{2}}(\mathbf{h})}}{\sigma_{0}^{2}}\right)\right], \\ E\left[\log\left(1 + \frac{h_{12}p_{12}(\mathbf{h})}{h_{12}p_{10}(\mathbf{h}) + \sigma_{2}^{2}}\right) + \log\left(1 + \frac{h_{21}p_{21}(\mathbf{h})}{h_{21}p_{20}(\mathbf{h}) + \sigma_{1}^{2}}\right) + \log\left(1 + \frac{h_{10}p_{10}(\mathbf{h}) + h_{20}p_{20}(\mathbf{h})}{\sigma_{0}^{2}}\right)\right]\right\} \end{split}$$

• Bounds not concave in power vector $\mathbf{p}(\mathbf{h}) = [p_{10}(\mathbf{h}) \ p_{12}(\mathbf{h}) \ p_{20}(\mathbf{h}) \ p_{21}(\mathbf{h}) \ p_{U_2}(\mathbf{h})]$

Properties of Sum-Rate-Optimal Power Allocation

Proposition 1 Let the effective channel gains normalized by the noise powers be defined as $s_{ij} = h_{ij}/\sigma_j^2$. Then, for the power control policy $\mathbf{p}^*(\mathbf{h})$ that maximizes the sum rate, we need

• $p_{10}^*(\mathbf{h}) = p_{20}^*(\mathbf{h}) = 0$, if $s_{12} > s_{10}$ and $s_{21} > s_{20}$

•
$$p_{10}^*(\mathbf{h}) = p_{21}^*(\mathbf{h}) = 0$$
, if $s_{12} > s_{10}$ and $s_{21} \le s_{20}$

•
$$p_{12}^*(\mathbf{h}) = p_{20}^*(\mathbf{h}) = 0$$
, if $s_{12} \le s_{10}$ and $s_{21} > s_{20}$

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$$oR$$

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$$oR$$

$$p_{12}^{*}(\mathbf{h}) = p_{20}^{*}(\mathbf{h}) = 0$$

$$if s_{12} \le s_{10} \text{ and } s_{21} \le s_{20}$$

Implications of the Optimal Power Allocation

- Block Markov superposition coding is simpler than originally thought.
 - Each transmitter either sends a cooperation signal or fresh information, but not both!
- The choice at each channel state "only" depends on the channel state.
 - Channel statistics, power constraints play no role in deciding which signals to transmit.
 - Except for the tiny little last case... which usually has very insignificant probability.
- The achivable rate expressions are greatly simplified, and are now concave.
- This simplified coding policy not only maximizes the sum rate, but also the individual rate constrains on R_1 and R_2 , and is optimal in terms of the entire rate region.
- Concave optimization problem over a convex constraint set, but non-differentiable.

Simplified Rate Region – Example

• Assume $s_{12} > s_{10}$, $s_{21} > s_{20}$ to illustrate the simplified rate region.

$$R_{1} < E \left[\log \left(1 + s_{12}p_{12}(\mathbf{h}) \right) \right]$$

$$R_{2} < E \left[\log \left(1 + s_{21}p_{21}(\mathbf{h}) \right) \right]$$

$$R_{1} + R_{2} < \min \left\{ E \left[\log \left(1 + s_{10}p_{1}(\mathbf{h}) + s_{20}p_{2}(\mathbf{h}) + 2\sqrt{s_{10}s_{20}p_{U_{1}}(\mathbf{h})p_{U_{2}}(\mathbf{h})} \right) \right],$$

$$E \left[\log \left(1 + s_{12}p_{12}(\mathbf{h}) \right) + \log \left(1 + s_{21}p_{21}(\mathbf{h}) \right) \right] \right\}$$

- Inequalities define either a pentagon like in the traditional MAC, or a triangle.
- All bounds concave in powers, and so is any weighted sum $\mu_1 R_1 + \mu_2 R_2$ at the corners.
- Sum rate not differentiable where the arguments of the min are equal.

Rate Maximization Using Subgradient Method

- Points on the rate region boundary can be obtained by maximizing $C_{\mu} = \mu_1 R_1 + \mu_2 R_2$.
- The optimization problem for arbitrary priorities μ_1 and μ_2 is given by

$$\max_{\mathbf{p}(\mathbf{h})} \mu_1 R_1 + \mu_2 R_2$$

s.t. $E_{3,4} [p_{10}(\mathbf{h})] + E_{1,2} [p_{12}(\mathbf{h})] + E [p_{U_1}(\mathbf{h})] \le \bar{p}_1$
 $E_{2,4} [p_{20}(\mathbf{h})] + E_{1,3} [p_{21}(\mathbf{h})] + E [p_{U_2}(\mathbf{h})] \le \bar{p}_2$

- $\{R_1, R_2\}$ is the corner of the pentagon obtained for a given power allocation policy.
- Gradient of the objective function does not exist everywhere, find subgradient g instead

$$C_{\boldsymbol{\mu}}(\mathbf{p}') \leq C_{\boldsymbol{\mu}}(\mathbf{p}) + (\mathbf{p}' - \mathbf{p})\mathbf{g}$$

• Use projected subgradient method to maximize C_{μ}

$$\mathbf{p}(k+1) = [\mathbf{p}(k) + \alpha_k \mathbf{g}_k]^+$$

• Provably converges for a diminishing stepsize α_k [Shor].

Convergence of the Projected Subgradient Algorithm



- Rate of convergence depends on the stepsize parameter.
- Subgradient method need not give a monotonically increasing function value.

Achievable Rate Region for Joint Power Control and User Cooperation



• Optimized power levels enlarge the achievable rate region significantly.

Summary and Conclusions

- Characterized the power control policies that are jointly optimal with Block Markov superposition coding.
- Using sub-gradient methods, obtained optimal power levels and corresponding rate region.
- Joint usage of cooperative diversity and time diversity: major improvements in capacity.
- Encoding and decoding is significantly simplified.
 - Transmitters send either cooperation or fresh information signals, but noth both.
- Optimal power policies also dictate MAC and routing policies
 - Cross layer design.

The Three User Cooperative Multiple Access Channel



 $Y_0 = h_{10}X_1 + h_{20}X_2 + h_{30}X_3 + N_0$ $Y_1 = h_{21}X_2 + h_{31}X_3 + N_1$ $Y_2 = h_{12}X_1 + h_{32}X_3 + N_2$ $Y_3 = h_{13}X_1 + h_{23}X_2 + N_3$

• Joint work with Cagatay Edemen



• Achievable rates obtained by block Markov Encoding [Sendonaris-Erkip-Aazhang 2003]



• Multiple Access Relay Channel [Sankaranarayanan, Kramer, Mandayam 2004]



 $Y_0 = h_{10}X_1 + h_{20}X_2 + h_{30}X_3 + N_0$ $Y_2 = h_{12}X_1 + N_2$ $Y_3 = h_{13}X_1 + N_3$

• Multiple Relay Channel [Schein, Gallager 00], [Kramer, Gastpar, Gupta 03], [Xie, Kumar 05]



• Multiple users mutually cooperate: fairer scheme, potentially higher rates, more ad-hoc.

Block Markov Encoding - Two Users

• Two user cooperation: each user's message is divided into two sub-messages

$$- w_1 = (w_{10}, w_{12}), \quad w_2 = (w_{20}, w_{21})$$

- Block Markov superposition coding
 - Build common information (X_{12}, X_{21})
 - Cooperatively send (U)
 - Inject new information (X_{10}, X_{20})

$$X_{1} = \sqrt{p_{10}(\mathbf{h})} X_{10} + \sqrt{p_{12}(\mathbf{h})} X_{12} + \sqrt{p_{u1}(\mathbf{h})} U$$
$$X_{2} = \sqrt{p_{20}(\mathbf{h})} X_{20} + \sqrt{p_{21}(\mathbf{h})} X_{21} + \sqrt{p_{u2}(\mathbf{h})} U$$

• When cooperative links stronger than direct links, should never send w_{i0} [Kaya-Ulukus 07]

Block Markov Encoding- Three Users

• Extension of Block Markov encoding to three user cooperation:

 $- w_1 = (w_{10}, w_{12}, w_{13}), \quad w_2 = (w_{20}, w_{21}, w_{23}), \quad w_3 = (w_{30}, w_{31}, w_{32})$

- Which messages should be decoded by which users?
 - Potentially too much interference.
- How should the cooperative signals be formed?
 - Should the users cooperate in pairs? Collectively?
- We propose a channel adaptive encoding/decoding approach.
- Drop w_{i0} for simplicity: assume stronger inter-user links as in two user case.



User	Own Messages	Decoded Messages
1	<i>w</i> ₁₂ , <i>w</i> ₁₃	
2	<i>w</i> ₂₁ , <i>w</i> ₂₃	
3	<i>w</i> ₃₁ , <i>w</i> ₃₂	



- Normalization: $s_{ij} = h_{ij}/\sigma_j^2$
- Assumption:

$$- s_{ij} > s_{i0}, \forall i, j \in \{1, 2, 3\}, i \neq j$$

 $- s_{12} > s_{13}.$

User	Own Messages	Decoded Messages
1	<i>w</i> ₁₂ , <i>w</i> ₁₃	
2	<i>w</i> ₂₁ , <i>w</i> ₂₃	<i>w</i> ₁₂ , <i>w</i> ₁₃
3	<i>w</i> ₃₁ , <i>w</i> ₃₂	<i>w</i> ₁₃



- Normalization: $s_{ij} = h_{ij}/\sigma_j^2$
- Assumption:
 - $s_{ij} > s_{i0}, \ \forall i, j \in \{1, 2, 3\}, \ i \neq j$
 - $s_{12} > s_{13},$
 - $s_{21} > s_{23}$

User	Own Messages	Decoded Messages
1	<i>w</i> ₁₂ , <i>w</i> ₁₃	<i>w</i> ₂₁ , <i>w</i> ₂₃
2	<i>w</i> ₂₁ , <i>w</i> ₂₃	w_{12}, w_{13}
3	<i>w</i> ₃₁ , <i>w</i> ₃₂	<i>w</i> ₁₃ , <i>w</i> ₂₃



- Normalization: $s_{ij} = h_{ij}/\sigma_j^2$
- Assumption:
 - $s_{ij} > s_{i0}, \ \forall i, j \in \{1, 2, 3\}, \ i \neq j$
 - $s_{12} > s_{13},$
 - $s_{21} > s_{23}$

$$-s_{32} > s_{31}$$

User	Own Messages	Decoded Messages
1	<i>w</i> ₁₂ , <i>w</i> ₁₃	<i>w</i> ₂₁ , <i>w</i> ₂₃ , <i>w</i> ₃₁
2	<i>w</i> ₂₁ , <i>w</i> ₂₃	<i>w</i> ₁₂ , <i>w</i> ₁₃ , <i>w</i> ₃₂ , <i>w</i> ₃₁
3	<i>w</i> ₃₁ , <i>w</i> ₃₂	<i>w</i> ₁₃ , <i>w</i> ₂₃

Block Markov Coding

User	Transmitted Codeword
1	$U(w'_{13}, w'_{23}, w'_{31}), U_1(w'_{12}, w'_{21}, U),$
	$X_{12}(w_{12}, U_1, U), X_{13}(w_{13}, U_1, U), X_{10}(w_{10}, X_{12}, X_{13}, U_1, U)$
2	$U(w'_{13}, w'_{23}, w'_{31}), U_1(w'_{12}, w'_{21}, U), U_3(w'_{32}, U),$
	$X_{21}(w_{21}, U_1, U_3, U), X_{23}(w_{23}, U_1, U_3, U), X_{20}(w_{20}, X_{21}, X_{23}, U_1, U_3, U)$
3	$U(w'_{13}, w'_{23}, w'_{31}), U_3(w'_{32}, U),$
	$X_{31}(w_{31}, U_3, U), X_{32}(w_{32}, U_3, U), X_{30}(w_{30}, X_{31}, X_{32}, U_3, U)$

$$X_{1} = \sqrt{P_{10}}X_{10} + \sqrt{P_{12}}X_{12} + \sqrt{P_{13}}X_{13} + \sqrt{P_{1U_{1}}}U_{1} + \sqrt{P_{1U}}U$$

$$X_{2} = \sqrt{P_{20}}X_{20} + \sqrt{P_{21}}X_{21} + \sqrt{P_{23}}X_{23} + \sqrt{P_{2U_{1}}}U_{1} + \sqrt{P_{2U_{3}}}U_{3} + \sqrt{P_{1U}}U$$

$$X_{3} = \sqrt{P_{30}}X_{30} + \sqrt{P_{31}}X_{31} + \sqrt{P_{32}}X_{32} + \sqrt{P_{3U_{3}}}U_{3} + \sqrt{P_{3U}}U$$

$$P_{10} + P_{12} + P_{13} + P_{1U_1} + P_{1U} \le P_1$$

$$P_{20} + P_{21} + P_{23} + P_{2U_1} + P_{2U_3} + P_{2U} \le P_2$$

$$P_{30} + P_{31} + P_{32} + P_{3U_3} + P_{3U} \le P_3$$

Rate Constraints for Error Free Decoding at User 1



- User 1 can decode w_{21} , w_{23} and w_{31} without error
- X_{32} and its cooperative version $U_3(w'_{32})$ are treated as noise at User 1

$$-A = s_{21}P_{2U_3} + s_{31}(P_{32} + P_{3U_3}) + 2\sqrt{s_{21}s_{31}P_{2U_3}P_{3U_3}} + 1$$

Rate Constraints for Error Free Decoding at User 2



$$\begin{split} R_{12} &< E \left[\log \left(1 + s_{12} P_{12} \right) \right] \\ R_{13} &< E \left[\log \left(1 + s_{12} P_{13} \right) \right] \\ R_{31} &< E \left[\log \left(1 + s_{32} P_{31} \right) \right] \\ R_{32} &< E \left[\log \left(1 + s_{32} P_{32} \right) \right] \\ R_1 &< E \left[\log \left(1 + s_{12} (P_{12} + P_{13}) \right) \right] \\ R_{12} + R_{31} &< E \left[\log \left(1 + s_{12} P_{12} + s_{32} P_{31} \right) \right] \\ R_{12} + R_{32} &< E \left[\log \left(1 + s_{12} P_{12} + s_{32} P_{32} \right) \right] \end{split}$$

- User 2 can decode all transmitted signals.
- MAC capacity region with 4 independent messages,
 - no interference terms.

Rate Constraints for Error Free Decoding at User 2 (cnt'd)



$$\begin{split} R_{13} + R_{31} &< E \left[\log \left(1 + s_{12}P_{13} + s_{32}P_{31} \right) \right] \\ R_{13} + R_{32} &< E \left[\log \left(1 + s_{12}P_{13} + s_{32}P_{32} \right) \right] \\ R_3 &< E \left[\log \left(1 + s_{32}(P_{31} + P_{32}) \right) \right] \\ R_1 + R_{31} &< E \left[\log \left(1 + s_{12}(P_{12} + P_{13}) + s_{32}P_{31} \right) \right] \\ R_1 + R_{32} &< E \left[\log \left(1 + s_{12}(P_{12} + P_{13}) + s_{32}P_{32} \right) \right] \\ R_{12} + R_3 &< E \left[\log \left(1 + s_{12}P_{12} + s_{32}(P_{31} + P_{32}) \right) \right] \\ R_{13} + R_3 &< E \left[\log \left(1 + s_{12}P_{13} + s_{32}(P_{31} + P_{32}) \right) \right] \\ R_1 + R_3 &< E \left[\log \left(1 + s_{12}(P_{12} + P_{13}) + s_{32}(P_{31} + P_{32}) \right) \right] \\ R_1 + R_3 &< E \left[\log \left(1 + s_{12}(P_{12} + P_{13}) + s_{32}(P_{31} + P_{32}) \right) \right] \end{split}$$

- User 2 can decode all transmitted signals.
- MAC capacity region with 4 independent messages,
 - no interference terms.



- User 3 can decode w_{13} and w_{23} without error
- X_{12}, X_{21} and their cooperative version $U_1(w'_{12}, w'_{21})$ are treated as noise component at User 3

$$-B = s_{13}(P_{12} + P_{1U_1}) + s_{23}(P_{21} + P_{2U_1}) + 2\sqrt{s_{13}s_{23}P_{1U_1}P_{2U_1}} + 1$$









$$\begin{split} R_{32} &< E \left[\log \left(1 + s_{20} P_{2U_3} + s_{30} (P_{32} + P_{3U_3}) + 2 \sqrt{s_{20} s_{30} P_{2U_3} P_{3U_3}} \right) \right] \\ R_{12} + R_{21} &< E \left[\log \left(1 + s_{10} (P_{12} + P_{1U_1}) + s_{20} (P_{21} + P_{2U_1}) + 2 \sqrt{s_{10} s_{20} P_{1U_1} P_{2U_1}} \right) \right] \\ R_{13} + R_{23} + R_{31} &< E \left[\log \left(1 + s_{10} (P_{13} + P_{1U}) + s_{20} (P_{23} + P_{2U}) + s_{30} (P_{31} + P_{3U}) \right. \\ &+ 2 (\sqrt{s_{10} s_{20} P_{1U} P_{2U}} + \sqrt{s_{10} s_{30} P_{1U} P_{3U}} + \sqrt{s_{20} s_{30} P_{2U} P_{3U}}) \right) \right] \\ R_{12} + R_{21} + R_{32} &< E \left[\log \left(1 + s_{10} (P_{12} + P_{1U_1}) + s_{20} (P_{21} + P_{2U_1} + P_{2U_3}) + s_{30} (P_{32} + P_{3U_3}) \right. \\ &+ 2 \sqrt{s_{10} s_{20} P_{1U_1} P_{2U_1}} + 2 \sqrt{s_{20} s_{30} P_{2U_3} P_{3U_3}} \right) \right] \\ R_{13} + R_{23} + R_{3} &< E \left[\log \left(1 + s_{10} (P_{13} + P_{1U}) + s_{20} (P_{23} + P_{2U} + P_{2U_3}) + s_{30} P_{3} \right. \\ &+ 2 \sqrt{s_{20} s_{30} P_{2U_3} P_{3U_3}} + 2 (\sqrt{s_{10} s_{20} P_{1U} P_{2U}} + \sqrt{s_{10} s_{30} P_{1U} P_{3U}} + \sqrt{s_{20} s_{30} P_{2U} P_{3U}}) \right) \right] \\ R_{1} + R_{2} + R_{31} < E \left[\log \left(1 + s_{10} P_{1} + s_{20} (P_{21} + P_{23} + P_{2U} + P_{2U_1}) + s_{30} (P_{31} + P_{3U}) \right. \\ &+ 2 \sqrt{s_{10} s_{20} P_{1U_1} P_{2U_1}} + 2 (\sqrt{s_{10} s_{20} P_{1U} P_{2U}} + \sqrt{s_{10} s_{30} P_{1U} P_{3U}} + \sqrt{s_{20} s_{30} P_{2U} P_{3U}}) \right) \right] \\ R_{1} + R_{2} + R_{3} < E \left[\log \left(1 + s_{10} P_{1} + s_{20} (P_{21} + P_{23} + P_{2U} + P_{2U_1}) + s_{30} (P_{31} + P_{3U}) \right. \\ &+ 2 \sqrt{s_{10} s_{20} P_{1U_1} P_{2U_1}} + 2 (\sqrt{s_{10} s_{20} P_{1U} P_{2U}} + \sqrt{s_{10} s_{30} P_{1U} P_{3U}} + \sqrt{s_{20} s_{30} P_{2U} P_{3U}}) \right) \right] \\ R_{1} + R_{2} + R_{3} < E \left[\log \left(1 + s_{10} P_{1} + s_{20} P_{2} + s_{30} P_{3} \right. \\ &+ 2 \sqrt{s_{10} s_{20} P_{1U_1} P_{2U_1}} + 2 \sqrt{s_{20} s_{30} P_{2U_3} P_{3U_3}} + 2 (\sqrt{s_{10} s_{20} P_{1U} P_{2U}} + \sqrt{s_{10} s_{30} P_{1U} P_{3U}} + \sqrt{s_{20} s_{30} P_{2U} P_{3U}}) \right) \right] \\ R_{1} + R_{2} + R_{3} < E \left[\log \left(1 + s_{10} P_{1} + s_{20} P_{2} + s_{30} P_{3} \right] \right]$$













Channel Ordering Assumption Revisited

• Considered channel orderings of type

$$s_{ij} > s_{ik}, \quad s_{ji} > s_{jk}, \quad s_{kj} > s_{ki}, \quad i \neq j \neq k$$

- Concept of "strong" and "weak" users
- Other channel orderings possible

 $s_{ij} > s_{ik}, \quad s_{jk} > s_{ji}, \quad s_{ki} > s_{kj}, \quad i \neq j \neq k$

- We need to update the encoding/decoding accordingly.
- Not asymmetric as before, no particular "strong user": strategy becomes somewhat different.
- Idea: choose the right encoding strategy based on the instantaneous channel ordering.



User	Own Messages	Decoded Messages
1	<i>w</i> ₁₂ , <i>w</i> ₁₃	
2	<i>w</i> ₂₁ , <i>w</i> ₂₃	
3	<i>w</i> ₃₁ , <i>w</i> ₃₂	



- Normalization: $s_{ij} = h_{ij}/\sigma_j^2$
- Assumption:

$$- s_{ij} > s_{i0}, \forall i, j \in \{1, 2, 3\}, i \neq j$$

 $- s_{12} > s_{13}.$

User	Own Messages	Decoded Messages
1	<i>W</i> ₁₂ , <i>W</i> ₁₃	
2	<i>w</i> ₂₁ , <i>w</i> ₂₃	<i>w</i> ₁₂ , <i>w</i> ₁₃
3	<i>w</i> ₃₁ , <i>w</i> ₃₂	<i>w</i> ₁₃



- Normalization: $s_{ij} = h_{ij}/\sigma_j^2$
- Assumption:
 - $s_{ij} > s_{i0}, \ \forall i, j \in \{1, 2, 3\}, \ i \neq j$
 - $s_{12} > s_{13},$
 - $s_{23} > s_{21}$

User	Own Messages	Decoded Messages
1	<i>w</i> ₁₂ , <i>w</i> ₁₃	<i>w</i> ₂₁
2	<i>w</i> ₂₁ , <i>w</i> ₂₃	w_{12}, w_{13}
3	<i>w</i> ₃₁ , <i>w</i> ₃₂	<i>w</i> ₁₃ , <i>w</i> ₂₃ , <i>w</i> ₂₁



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$$-s_{31} > s_{32}$$

User	Own Messages	Decoded Messages
1	<i>w</i> ₁₂ , <i>w</i> ₁₃	<i>w</i> ₂₁ , <i>w</i> ₃₁ , <i>w</i> ₃₂
2	<i>w</i> ₂₁ , <i>w</i> ₂₃	<i>w</i> ₁₂ , <i>w</i> ₁₃ , <i>w</i> ₃₂
3	<i>w</i> ₃₁ , <i>w</i> ₃₂	<i>w</i> ₁₃ , <i>w</i> ₂₃ , <i>w</i> ₂₁

Block Markov Coding

User	Transmitted Codeword
1	$U(w'_{13}, w'_{21}, w'_{32}), U_1(w'_{12}, U), U_3(w'_{31}, U),$
	$X_{12}(w_{12}, U_1, U), X_{13}(w_{13}, U_3, U), X_{10}(w_{10}, X_{12}, X_{13}, U_1, U_3, U)$
2	$U(w'_{13}, w'_{21}, w'_{32}), U_1(w'_{12}, U), U_2(w'_{23}, U),$
	$X_{21}(w_{21}, U_1, U), X_{23}(w_{23}, U_2, U), X_{20}(w_{20}, X_{21}, X_{23}, U_1, U_2, U)$
3	$U(w'_{13}, w'_{21}, w'_{32}), U_3(w'_{31}, U), U_2(w'_{23}, U),$
	$X_{31}(w_{31}, U_3, U), X_{32}(w_{32}, U_2, U), X_{30}(w_{30}, X_{31}, X_{32}, U_2, U_3, U)$

$$\begin{aligned} X_1 &= \sqrt{P_{10}} X_{10} + \sqrt{P_{12}} X_{12} + \sqrt{P_{13}} X_{13} + \sqrt{P_{1U_1}} U_1 + \sqrt{P_{1U_3}} U_3 + \sqrt{P_{1U}} U \\ X_2 &= \sqrt{P_{20}} X_{20} + \sqrt{P_{21}} X_{21} + \sqrt{P_{23}} X_{23} + \sqrt{P_{2U_1}} U_1 + \sqrt{P_{2U_2}} U_2 + \sqrt{P_{1U}} U \\ X_3 &= \sqrt{P_{30}} X_{30} + \sqrt{P_{31}} X_{31} + \sqrt{P_{32}} X_{32} + \sqrt{P_{3U_2}} U_2 + \sqrt{P_{3U_3}} U_3 + \sqrt{P_{3U}} U \end{aligned}$$

 $P_{10} + P_{12} + P_{13} + P_{1U_1} + P_{1U_3} + P_{1U} \le P_1$ $P_{20} + P_{21} + P_{23} + P_{2U_1} + P_{2U_2} + P_{2U} \le P_2$ $P_{30} + P_{31} + P_{32} + P_{3U_2} + P_{3U_3} + P_{3U} \le P_3$



- User 1 can decode w_{21} , w_{31} and w_{32} without error
- $U_2(w'_{23})$ is treated as noise at User 1

$$-A = 1 + s_{21}P_{2U2} + s_{31}P_{3U2} + 2\sqrt{s_{21}s_{31}P_{2U2}P_{3U2}}$$



- User 2 can decode w_{12} , w_{13} and w_{32} without error
- $U_3(w'_{31})$ is treated as noise at User 2

$$-B = 1 + s_{12}P_{1U3} + s_{32}P_{3U3} + 2\sqrt{s_{12}s_{32}P_{1U3}P_{3U3}}$$



- User 3 can decode w_{13} , w_{21} and w_{23} without error
- $U_1(w'_{12})$ is treated as noise at User 3

$$-C = 1 + s_{13}P_{1U1} + s_{23}P_{2U1} + 2\sqrt{s_{13}s_{23}P_{1U1}P_{2U1}}$$











$$\begin{split} & R_{12} < E \left[\log \left(1 + s_{10}(P_{12} + P_{1U1}) + s_{20}(P_{21} + P_{2U1}) + 2\sqrt{s_{10}s_{20}P_{1U1}P_{2U1}} \right) \right] \\ & R_{31} < E \left[\log \left(1 + s_{10}(P_{13} + P_{1U3}) + s_{30}(P_{31} + P_{3U3}) + 2\sqrt{s_{10}s_{30}P_{1U3}P_{3U3}} \right) \right] \\ & R_{23} < E \left[\log \left(1 + s_{20}(P_{23} + P_{2U2}) + s_{30}(P_{32} + P_{3U2}) + 2\sqrt{s_{20}s_{30}P_{2U2}P_{3U2}} \right) \right] \\ & R_{12} + R_{23} < E \left[\log \left(1 + s_{10}(P_{12} + P_{1U1}) + s_{20}(P_{21} + P_{23} + P_{2U1} + P_{2U2}) + s_{30}(P_{32} + P_{3U2}) + 2\sqrt{s_{10}s_{20}P_{1U1}P_{2U1}} + 2\sqrt{s_{20}s_{30}P_{2U2}P_{3U2}} \right) \right] \\ & R_{12} + R_{31} < E \left[\log \left(1 + s_{10}(P_{12} + P_{13} + P_{1U1} + P_{1U3}) + s_{20}(P_{21} + P_{2U1}) + s_{30}(P_{31} + P_{3U3}) + 2\sqrt{s_{10}s_{20}P_{1U1}P_{2U1}} + 2\sqrt{s_{10}s_{30}P_{1U3}P_{3U3}} \right) \right] \\ & R_{23} + R_{31} < E \left[\log \left(1 + s_{10}(P_{13} + P_{1U3}) + s_{20}(P_{23} + P_{2U2}) + s_{30}(P_{31} + P_{32} + P_{3U2} + P_{3U3}) + 2\sqrt{s_{10}s_{30}P_{1U3}P_{3U3}} + 2\sqrt{s_{20}s_{30}P_{2U2}P_{3U2}} \right) \right] \\ & R_{12} + R_{23} + R_{31} < E \left[\log \left(1 + s_{10}(P_{12} + P_{13} + P_{1U1} + P_{1U3}) + s_{20}(P_{21} + P_{23} + P_{2U1} + P_{2U2}) + s_{30}(P_{31} + P_{32} + P_{3U2} + P_{3U2} + P_{3U3}) + 2\sqrt{s_{10}s_{20}P_{1U1}P_{2U1}} + 2\sqrt{s_{10}s_{30}P_{1U3}P_{3U3}} + 2\sqrt{s_{20}s_{30}P_{2U2}P_{3U2}} \right) \right] \\ & R_{1} + R_{2} + R_{3} < E \left[\log \left(1 + s_{10}(P_{12} + P_{13} + P_{1U1} + P_{1U3} \right) + s_{20}(P_{21} + P_{23} + P_{2U1} + P_{2U2}) + s_{30}(P_{31} + P_{32} + P_{3U2} + P_{3U3}) + 2\sqrt{s_{10}s_{20}P_{1U1}P_{2U1}} + 2\sqrt{s_{10}s_{30}P_{1U3}P_{3U3}} + 2\sqrt{s_{20}s_{30}P_{2U2}P_{3U2}} \right) \right] \\ & R_{1} + R_{2} + R_{3} < E \left[\log \left(1 + s_{10}P_{1} + s_{20}P_{2} + s_{30}P_{3} + 2\sqrt{s_{20}s_{30}P_{2U2}P_{3U2}} \right) + 2\sqrt{s_{10}s_{30}P_{1U}P_{3U}} + \sqrt{s_{20}s_{30}P_{2U}P_{3U2}} \right) \right] \\ & + 2\sqrt{s_{10}s_{30}P_{1U3}P_{3U3}} + 2\sqrt{s_{20}s_{30}P_{2U}P_{3U2}} + 2(\sqrt{s_{10}s_{30}P_{1U}P_{2U}} + \sqrt{s_{10}s_{30}P_{1U}P_{3U}} + \sqrt{s_{20}s_{30}P_{2U}P_{3U}}) \right) \\ \end{array}$$

Achievable Rates for Three User Cooperation - Channel Ordering I



Achievable Rates for Three User Cooperation - Channel Ordering II





Summary and Conclusions

- Proposed a new block Markov superposition type encoding policy for the three user MAC
 - Non-trivial generalization of the two user policy
 - Channel adaptive
- Obtained the achievable rate regions
- Significant improvement with respect to two user cooperation
 - Multi-user cooperation quite promising as a means of improving diversity.
- Adapting the encoding, decoding and transmit strategies to the channel in cooperative networks is a key approach for improving achievable rates.